The Effect of Saturation on Belief Propagation Decoding of LDPC Codes

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Abstract—We consider the effect of LLR saturation on belief propagation decoding of low-density parity-check codes. Saturation occurs universally in practice and is known to have a significant effect on error floor performance. Our focus is on threshold analysis and stability of density evolution.

We analyze the decoder for certain low-density parity-check code ensembles and show that belief propagation decoding generally degrades gracefully with saturation. Stability of density evolution is, on the other hand, rather strongly affected by saturation and the asymptotic qualitative effect of saturation is similar to reduction of variable node degree by one.

I. INTRODUCTION

Standard belief propagation (BP) decoding of binary LDPC codes involves passing messages typically representing log-likelihood ratios (LLRs) which can take any value in $\overline{\mathbb{R}} \triangleq \mathbb{R} \cup \{\pm\infty\}$ [1]. Practical implementations of decoders typically use uniformly quantized and bound LLRs. Hence, it is of interest to understand the effect of saturation of LLR magnitudes as a perturbation of full belief propagation. We call such a clipped or a saturated decoder as a *saturating* belief propagation decoder (SatBP). Note that the decoder is strictly speaking not a BP decoder, but we adhere to the BP nomenclature as we view SatBP as a perturbation of BP.

The papers [2]-[5] consider the effect of saturation on error floor performance. It is observed in these works that saturation can limit the ability of decoding to escape trapping set behavior, thereby worsening error floor performance. Although we take a different approach in this paper by focussing on asymptotic behavior, the underlying message is similar: saturation can dramatically effect the stability of the decoder. In [6], [7] some decoder variations are given the help reduce error floors. Here we see an explicit effort to ameliorate the effect of saturation. A related but distinct direction was taken in [8]. There the authors made modifications to discrete node update rules so as to reduce error floor failure events. There have been other work that examine the effects of practical concessions. In [9] the authors consider the effect of quantization in LDPC coded flash memories. In [10], [11] model the effect of saturation and quantization as noise terms. Finally, in [12] an analysis was done to evaluate the effect on capacity on quantization of channel outputs.

In the design of capacity-achieving codes it would be helpful to understand how practical decoder concessions, like saturation, affect performance. In particular, if LLRs are saturated at magnitude K then how much degradation from the BP threshold should be expected. Naturally, one expects that as $K \to +\infty$, that one can reliably transmit arbitrarily close

to the BP threshold [1]. We will see that this is not entirely correct and that, in particular, saturation can undermine the stability of the perfect decoding fixed point.

II. BP DECODING, DENSITY EVOLUTION AND THE WASSERSTEIN DISTANCE

In this section we briefly review the BP decoder and the density evolution analysis [13] in the case of transmission over a general BMS channel. Most of the material presented here can be found in [1].

Let $X(=\pm 1)$ denote the channel input, let Y denote the channel output, and let $p(Y=y\,|\,X=x)$ denote the *transition probability* of the channel. We generally characterize the channel by its so-called L-density, c which is the distribution of $\ln\frac{p(Y\,|\,X=1)}{p(Y\,|\,X=-1)}$ conditioned on X=1. Generally, we may assume that $Y=\ln\frac{p(Y\,|\,X=1)}{p(Y\,|\,X=-1)}$, i.e., the output of the channel is the associated LLR. Channel symmetry is the condition $p(Y=y\,|\,X=x)=p(Y=-y\,|\,X=-x)$ and the L-densities c that result are symmetric, [1], which means $e^{-\frac{1}{2}x}c(x)$ is an even function of x. We recall that all densities which stem from BMS channels are symmetric, see [1, Sections 4.1.4, 4.1.8 and 4.1.9].

Given Z distributed according to \mathfrak{c} , we write \mathfrak{c} to denote the distribution of $\tanh(Z/2)$, and $|\mathfrak{c}|$ to denote the distribution of $|\tanh(Z/2)|$. We refer to these as the D and |D| distributions. We use $|\mathfrak{C}|$ to denote the corresponding cumulative |D| distribution, see [1, Section 4.1.4]. Under symmetry the distribution of |Z| determines the distribution of Z.

For threshold analysis of LDPC ensembles we typically consider a parameterized family of channels and write $\{BMS(\sigma)\}$ to denote the family as parameterized by the scalar σ . Often it will be more convenient to denote this family by $\{c_{\sigma}\}$, i.e., to use the family of L-densities which characterize the channel family. One natural candidate for the parameter σ is the entropy of the channel denoted by h. Thus, we also consider the characterization of the family given by the BMS(h).

Let $p_{Z\,|\,X}(z\,|\,x)$ denote the transition probability associated to a BMS channel c' and let $p_{Y\,|\,X}(y\,|\,x)$ denote the transition probability of another BMS channel c. We then say that c' is *degraded* with respect to c, denoted c \prec c', if there exists a channel $p_{Z\,|\,Y}(z\,|\,y)$ so that $p_{Z\,|\,X}(z\,|\,x) = \sum_y p_{Y\,|\,X}(y\,|\,x) p_{Z\,|\,Y}(z\,|\,y)$.

A BMS channel family $\{BMS(\sigma)\}\frac{\overline{\sigma}}{\underline{\sigma}}$ is said to be *ordered* (by degradation) if $\sigma_1 \leq \sigma_2$ implies $c_{\sigma_1} \prec c_{\sigma_2}$.

Useful functionals of densities include the Battacharyya, the entropy, and the error probability functionals. For a density a,

these are denoted by $\mathfrak{B}(\mathsf{a})$, $\mathsf{H}(\mathsf{a})$, and $\mathfrak{E}(\mathsf{a})$, respectively and are defined by

$$\begin{split} \mathfrak{B}(\mathbf{a}) &= \mathbb{E}(e^{-Z/2}), \quad \mathbf{H}(\mathbf{a}) = \mathbb{E}(\log_2(1+e^{-Z})), \\ \mathfrak{E}(\mathbf{a}) &= \mathbb{P}\{Z<0\} + \frac{1}{2}\mathbb{P}\{Z=0\} \end{split}$$

where Z is distributed according to a. Note that these definitions are valid even if a is not symmetric, although they loose some of their original meaning. We will apply these definitions, especially \mathfrak{B} , to saturated densities that are not necessarily symmetric.

1) BP Decoder and Density Evolution: The definition of the standard BP decoder can be found in [1]. The asymptotic performance of the BP decoder is given by the density evolution technique [1], [13].

Definition 1 (Density Evolution for BP Decoder cf. [1]): For $\ell \geq 1$, the DE equation for a (d_l, d_r) -regular ensemble is given by

$$T(\mathsf{c},\mathsf{x}) \triangleq \mathsf{x}_{\ell} = \mathsf{c} \circledast (\mathsf{x}_{\ell-1}^{\boxtimes d_r-1})^{\circledast d_l-1}.$$

Here, c is the L-density of the BMS channel over which transmission takes place and x_ℓ is the density emitted by variable nodes in the ℓ -th round of density evolution. Initially we have $x_0 = \Delta_0$, the delta function at 0. The operators \circledast and \boxdot correspond to the convolution of densities at variable and check nodes, respectively, see [1, Section 4.1.4]. \blacksquare Discussion: The DE analysis is simplified when we consider the class of symmetric message-passing decoders. The definition of symmetric message-passing decoders can be found in [1]. Note that this definition of symmetry pertains to the actual messages in the decoder and not to the densities which appear in the DE analysis. We will see later that the clipped or the saturated decoder is a symmetric message-passing decoder in this sense and hence its DE analysis is simplified by restricting to consideration of the all-zero codeword.

Definition 2 (BP Threshold for regular ensembles): Consider an ordered and complete channel family $\{c_h\}$. Let $x_\ell(h)$ denote the distribution in the ℓ -th round of DE when the channel is c_h . Then the BP threshold of the (d_l,d_r) -regular ensemble is defined as $h^{\text{BP}}(d_l,d_r,\{c_h\}) = \sup\{h: x_\ell(h) \overset{\ell \to \infty}{\to} \Delta_{+\infty}\}$. Under symmetry an equivalent definition is

$$\mathbf{h}^{\mathrm{BP}}(d_l,d_r,\{\mathsf{c_h}\}) = \sup\{\mathbf{h}: \mathfrak{E}(\mathsf{x}_\ell(\mathsf{h})) \overset{\ell \to \infty}{\to} 0\}.$$

The later form is more convenient for our purposes and it is the one we shall adopt.

In the sequel we will use the Wasserstein metric to measure distance between distributions. We recall the definition of the Wasserstein metric below. For details see [14].

Definition 3 (Wasserstein Metric – [15, Chapter 6]): Let $|\mathfrak{a}|$ and $|\mathfrak{b}|$ denote two |D|-distributions. The Wasserstein metric, denoted by $d(|\mathfrak{a}|, |\mathfrak{b}|)$, is defined as

$$d(|\mathfrak{a}|,|\mathfrak{b}|) = \sup_{f(x) \in \operatorname{Lip}(1)[0,1]} \left| \int_0^1 f(x) (|\mathfrak{a}|(x) - |\mathfrak{b}|(x)) \, \mathrm{d}x \right|, \quad (1)$$

III. SATURATED BELIEF PROPAGATION DECODING

We first consider the analysis of the saturated BP decoder. More precisely, we consider decoding with BP rules but with messages restricted to the domain [-K, K] for some K > 0.

A. Saturated Decoder

Definition 4 (Saturation): We define the saturation operation at $\pm K$ for some $K \in \mathbb{R}^+$, denoted $|\cdot|_K$, by

$$\lfloor x \rfloor_K = \min(K, |x|) \cdot \operatorname{sgn}(x). \tag{2}$$

Definition 5 (SatBP Decoder): Consider the standard (d_l, d_r) -regular ensemble. The saturated BP decoder is defined by the following rules. Let $\phi^{(\ell)}(\mu_1, \dots, \mu_{d_r-1})$ and $\psi^{(\ell)}(\mu_1, \dots, \mu_{d_l-1})$ denote the outgoing message from the check node and the variable node side respectively. Abusing the notation above and denoting the incoming messages on both the check node and the variable node side by μ_i , we have

$$\phi^{(\ell)}(\mu_1, \dots, \mu_{d_r - 1}) = \left[2 \tanh^{-1} \left(\prod_{i=1}^{d_r - 1} \tanh(\mu_i / 2) \right) \right]_K,$$

$$\psi^{(\ell)}(\mu_1, \dots, \mu_{d_l - 1}) = \left[\mu_0 + \sum_{i=1}^{d_l - 1} \mu_i \right]_K,$$

where μ_0 is the message coming from the channel. Also, we set $\phi^{(0)}(\mu_1,\ldots,\mu_{d_r-1})=0$.

Using Definition 4.83 in [1] we have the following.

Lemma 6 (SatBP Decoder is symmetric): The SatBP decoder given in Definition 5 is a symmetric decoder.

Discussion: The symmetry of the message-passing decoder together with symmetry of the channel allows us to use the allzero codeword assumption. This along with the concentration results (see Theorem 4.94 in [1]) allows to write down the density evolution of the SatBP decoder in the usual way.

Given $X \sim a$ let $|a|_K$ denote the distribution of $|X|_K$. Note that the saturation operation can be viewed as a channel taking X to $|X|_K$. We immediately have $a \prec |a|_K$. In general $|a|_K$ will not be symmetric even if a is symmetric since we will not typically have $|\mathbf{a}|_K(-K) = e^{-K}|\mathbf{a}|_K(K)$. If a is symmetric then we will have $|a|_K(-K) \le e^{-K}|a|_K(K)$. Although using lemma 6 one can write down the DE recursion for the SatBP decoder, we know that in general the densities will not be symmetric. Two of the most useful properties of DE for BP is that it preserves both symmetry of densities and ordering by degradation. These properties are sacrificed by saturation, but can be recovered with a slight variation. The idea is to slightly degrade the density by moving some probability mass from K to -K. This can be interpreted operationally as flipping the sign of a message with magnitude K with some probability λ . The flipping rate λ is chosen so that the resulting probability that the sign of the message is incorrect is $e^{-K}/(1+e^{-K})$. In general λ is upper bounded by this value and for large K this is a small perturbation. With this perturbation both density symmetry and ordering by degradation are recovered. Let us introduce the notation D(p, z) to denote the density $D(p,z)=p\Delta_{-z}+(1-p)\Delta_z$. Using this notation we have for symmetric a, $\lfloor {\sf a} \rfloor_K=\gamma D(q,z)(x)+{\sf a}(x)\mathbb{1}_{\{|x|< K\}}$ where $\gamma=\mathbb{P}_{\sf a}\{|x|\geq K\}$ and $\gamma q=\mathbb{P}_{\sf a}\{x\leq -K\}$.

Lemma 7 (Symmetric Saturation): Given a symmetric density a we define $\lfloor \mathtt{a} \rfloor_{K_{\mathrm{sym}}} = \gamma D(p,z)(x) + \mathtt{a}(x) \mathbb{1}_{\{|x| < K\}}$ where $p = e^{-K}/(1+e^{-K})$ and $\gamma = \mathbb{P}_{\mathtt{a}}\{|x| \ge K\}$. Then, (i) $\lfloor \mathtt{a} \rfloor_{K_{\mathrm{sym}}}$ is a symmetric L-density and (ii) $\lfloor \mathtt{a} \rfloor_{K} \prec \lfloor \mathtt{a} \rfloor_{K_{\mathrm{sym}}}$.

Proof: Part (i) is immediate. To prove part (ii) we note that comparing with the unsymmetrized case we see that $p \geq q$. Thus, $\lfloor \mathbf{a} \rfloor_{K_{\mathrm{sym}}}$ can be realized by taking messages with distribution $\lfloor \mathbf{a} \rfloor_K$ and flipping the sign of a message with magnitude K by a quantity λ which is determined by $p = \frac{e^{-K}}{1+e^{-K}} = \lambda(1-q) + (1-\lambda)q$.

As a consequence of Lemma 7, we will term the operation used to obtain $\lfloor a \rfloor_{K_{\text{sym}}}$ from a as *symmetric-saturation*.

We summarize all the claims above in the following.

Corollary 8: We have $a \prec \lfloor a \rfloor_K \prec \lfloor a \rfloor_{K_{\text{sym}}}$.

Lemma 9: Let a be a symmetric L-density. Then, $d(\mathsf{a}, |\mathsf{a}|_{K_{\mathrm{sym}}}) \leq 1 - \tanh(K/2)$.

Proof: For any $0 \le z < K$ we have $\mathbb{P}_{\mathsf{a}}\{x \le z\} = \mathbb{P}_{\lfloor \mathsf{a} \rfloor_K}\{x \le z\} = \mathbb{P}_{\lfloor \mathsf{a} \rfloor_{K_{\mathrm{sym}}}}\{x \le z\}$ and for any $z \ge K$ we have $1 = \mathbb{P}_{\lfloor \mathsf{a} \rfloor_K, \mathsf{sym}}\{x \le z\} = \mathbb{P}_{\lfloor \mathsf{a} \rfloor_{K_{\mathrm{sym}}}}\{x \le z\}$. Since $\tanh(x/2)$ is increasing and $\tanh(-x/2) = -\tanh(x/2)$ we have $|\lfloor \mathfrak{A} \rfloor_{K_{\mathrm{sym}}}|(z) = \mathbb{I}_{\{z < \tanh(K/2)\}}|\mathfrak{A}|(z) + \mathbb{I}_{\{z \ge \tanh(K/2)\}}$. By [14], we have that the Wasserstein distance is equivalent to the L_1 norm of the difference between the |D|-distribtions. Clearly, the distance is bounded by $1 - \tanh(K/2)$.

Definition 10 (DE for Sym. and Non-Sym. Saturation): The DE for non-symmetric saturation decoder is $S_{K_{\text{sym}}}(\mathsf{c},\mathsf{x}) = \lfloor T(\mathsf{c},\mathsf{x}) \rfloor_{K_{\text{sym}}}$ and for the non-symmetric saturation decoder is $S_K(\mathsf{c},\mathsf{x}) = \lfloor T(\mathsf{c},\mathsf{x}) \rfloor_K$.

We now estimate the distance between the densities appearing in the DE of usual BP and the DE of the symmetric-saturation operation.

Lemma 11 (Distance Between Symmetric-SatBP and BP): Consider ℓ iterations of the forward DE for the usual BP and the Symmetric-Saturation operation. Then

$$\mathfrak{B}(S_{K_{\mathrm{sym}}}^{(\ell)}(\mathsf{c},\Delta_0)) \leq \mathfrak{B}(T^{(\ell)}(\mathsf{c},\Delta_0)) + 2\sqrt{2}e^{\frac{-K + \ell \cdot \ln(2(d_l - 1)(d_r - 1))}{2}}$$

Proof: Using the triangle inequality, (viii), Lem. 13 in [14] and Lemma 9 we obtain the upper bound $d(T^{(\ell)}(\mathsf{c},\Delta_0),S^{(\ell)}_{K_{\mathrm{sym}}}(\mathsf{c},\Delta_0)) \leq \alpha_\ell d(T^{(\ell-1)}(\mathsf{c},\Delta_0),S^{(\ell-1)}_{K_{\mathrm{sym}}}(\mathsf{c},\Delta_0)) + 1 - \tanh\left(\frac{K}{2}\right),$ where $\alpha_\ell = 2(d_l-1)\sum_{j=1}^{d_r-1}(1-\mathfrak{B}^2(\mathsf{a}))^{\frac{d_r-1-j}{2}}(1-\mathfrak{B}^2(\mathsf{b}))^{\frac{j-1}{2}},$ a = $T^{(\ell-1)}(\mathsf{c},\Delta_0)$ and b = $S^{(\ell-1)}_{K_{\mathrm{sym}}}(\mathsf{c},\Delta_0)$. Continuing with the above inequality we get the upper bound, $(1-\tanh\left(\frac{K}{2}\right))(1+\alpha_\ell+\alpha_\ell\alpha_{\ell-1}+\ldots+\alpha_\ell\alpha_{\ell-1}+\ldots+\alpha_\ell\alpha_{\ell-1}\cdots\alpha_2).$ In general, we are transmitting below the BP threshold, so both a and b could be close to $\Delta_{+\infty}$. Thus, $(1+\alpha_\ell+\alpha_\ell\alpha_{\ell-1}+\ldots+\alpha_\ell\alpha_{\ell-1}\cdots\alpha_2)\leq (2(d_l-1)(d_r-1))^\ell.$ Using $1-\tanh(K/2)\leq 2e^{-K}$ and (ix) Lemma 13 in [14] we get the lemma.

IV. CONVERGENCE OF NONSYMMETRIC SATURATED DE

In the previous section we show that, when transmitting below the threshold of the full BP decoder, the Battacharrya parameter of the densities in the symmetric SatBP decoder can be arbitrarily small. In this section we make the connection to the non-symmetric SatBP.

A. Non-symmetrized SatBP Decoder

We now show that the Battacharrya parameter for the non-symmetric SatBP decoder also can be made as small as desired by choosing K large enough. We first consider a fixed computation tree and then average over the tree ensemble.

We begin with an operational description of symmetrization. Consider a fixed tree T of depth ℓ . Let Y denote the vector of received values associated to the variable nodes under the all-zero codeword assumption. In addition, for each variable node we assume an independent random variable uniformly distributed on [0,1]. We denote the vector of these variables by Z. Now, the node operations correspond to BP except that outgoing messages from the variable nodes are magnitude saturated at K. Independent random variables are used for the flipping operation at each node, where the flipping probabilty is determined by density evolution. If the outgoing message has magnitude K then its sign is flipped if $Z_v < \lambda_v$ where λ_v is the appropriate flipping probability.

The distribution of the outgoing message z is $S_{K_{\text{sym}}}^{(\ell)}(\mathsf{c}, \Delta_0)$). Let us consider the conditional distribution $p(z \mid Y, Z)$. We obtain $S_{K_{\text{sym}}}^{(\ell)}(\mathsf{c}, \Delta_0)$ by averaging over Y, Z and the code ensemble. Let A_K denote the event that no flipping occurs. $p(z \mid Y) = p(z \mid Y, A_K)p(A_K) + p(z \mid Y, \bar{A}_K)(1 - p(A_K))$. Averaging over Y and re-writing we obtain

$$p(z \mid A_K) = (p(z) - p(z \mid \bar{A}_K)(1 - p(A_K)))/p(A_K).$$

Now $p(z \mid A_K)$ is the distribution of the non-symmetric SatBP decoder. Intuitively one expects $p(z \mid \bar{A}_K)$ to be inferior (higher probability of error, larger Battacharyya parameter) to $p(z \mid A_K)$, but this appears difficult to prove.

Let us compute the probability of A_K . Let the received LLR magnitude of a variable node v be $z \geq K$. The probability with which we flip the bit is such that the final error probability is equal to $\frac{e^{-K}}{1+e^{-K}}$. For received LLR magnitude of z, the probability that it is received correctly is $\frac{1}{1+e^{-z}}$. As a consequence we get, $\frac{e^{-K}}{1+e^{-K}} = \lambda_v \frac{1}{1+e^{-z}} + (1-\lambda_v) \frac{e^{-z}}{1+e^{-z}}$, where λ_v is the flipping probability of variable node v. Solving we get $\lambda_v = \frac{e^{-K}}{1+e^{-K}} \frac{1-e^{-z+K}}{1-e^{-z}} \leq \frac{e^{-K}}{1+e^{-K}}$. Thus the probability that a variable node, with a received LLR magnitude greater than K, is not flipped is at least equal to $\frac{1}{1+e^{-K}} \geq 1-e^{-K}$. Hence, we get $p(A_K) \geq (1-e^{-K})^{|V(\mathsf{T})|} \geq 1-e^{-K}|V(\mathsf{T})|$ where $|V(\mathsf{T})|$ is the number of variable nodes in the tree. From the above analysis we have the following lemma.

Lemma 12 (SatBP Decoder versus Symmetrized SatBP): For any $\epsilon>0$ and $\ell\in\mathbb{N}$, there exists a K large enough such that $\mathfrak{B}(S_K^{(\ell)}(\mathsf{c},\Delta_0))\leq \frac{1}{1-\epsilon}\,\mathfrak{B}(S_{K_{\mathrm{sym}}}^{(\ell)}(\mathsf{c},\Delta_0)).$

V. STABILITY ANALYSIS

An important part of the asymptotic analysis of LDPC codes involves the analysis of the convergence of DE to a zero error state. In this section we analyze the stability of the SatBP. We begin with some necessary conditions.

For stability of the zero error condition there must exist a positive invariant set of zero error distributions, i.e, a subset S of distributions so that $\mathfrak{E}(\mathsf{s}) = 0$ for all $\mathsf{s} \in S$ and $S_K(\mathsf{c},\mathsf{s}) \in S$. Existence of S follows easily from the compactness of the space of densities and continuity of DE.

Lemma 13: Assume the channel c has support at -L, L > 0. In an irregular ensemble with minimum variable degree d_l the support of all densities in S must lie in $[L/(d_l-2),\infty)$.

Proof: Assume $\mathsf{a}^{(0)} \in \mathcal{S}$ has support at $z < L/(d_l-2)$. Then $\mathsf{b}^{(0)}$ has support on $[-\infty,z]$. Hence $\mathsf{a}^{(1)}$ has support on $[-\infty,z_1]$ where $z_1=(d_l-1)z-L< L-2\delta$ where $\delta=L-(d_l-2)z$. By induction we have $\mathsf{a}^{(k)}$ has support on $[-\infty,L-2^k\delta]$ and for k large enough the probability of error is positive.

A. Failure of Stability with Degree Two

From Lemma 13 we have immediately

Lemma 14: In an irregular ensemble with $\lambda_2 > 0$ no invariant S exists for any value of $K < \infty$ unless the channel is the BEC.

Proof: If $d_l = 2$ and the channel is not the BEC and hence has support on $(-\infty, 0)$, then Lemma 13 shows that there can be no positive invariant zero-error set of distributions with support on [-K, K].

In the case of the BEC it can be seen that saturated DE matches unsaturated DE except that the mass at $+\infty$ in unsaturated DE is not placed at +K. Hence, stability is unaffected by saturation.

If the channel has infinite support, then there is no possibility of stability under saturation. The condition on the finite channel support is given later in the section on stability with degree ≥ 3 .

B. Near Stability

The stability analysis of standard irregular ensembles under BP decoding rests on on the results $\mathfrak{B}(\mathsf{c}\circledast\lambda(\mathsf{a}))=\mathfrak{B}(\mathsf{c})\lambda(\mathfrak{B}(\mathsf{a}))$ and $\mathfrak{B}(\rho(\mathsf{a}))\leq 1-\rho(1-\mathfrak{B}(\mathsf{a}))$. The first equation continues to hold without symmetry of a or c . The check node inequality, however, does not necessarily hold without symmetry. We have, however, a substitute with a slight variation on a result from [16]: for any L-densities a and b we have $\mathfrak{B}(\mathsf{a}\boxtimes\mathsf{b})\leq \mathfrak{B}(\mathsf{a})+\mathfrak{B}(\mathsf{b})$. This result holds for a wide range of check node update operations including BP and the min-sum decoder. To incorporate saturation into the analysis based on the Bhattacharyya parameter we have the inequality $\mathfrak{B}(\lfloor \mathsf{a} \rfloor_K) \leq \mathfrak{B}(\mathsf{a}) + e^{-K/2}$. Note that because of saturation, the minimum value of the Battacharyya parameter is equal to $e^{-K/2}$.

Minimum variable node degree equal to 2: Let $a^{(k)}$ be any L-density which need not be symmetric. Consider an irregular ensemble and assume $\rho'(1) \mathfrak{B}(a^{(k)}) < 1$. Using

the notation $\bar{\lambda}_2=1-\lambda_2$ we upper bound $\mathfrak{B}(\mathsf{a}^{(k+1)})$ for $\mathsf{a}^{(k+1)}=S_K(\mathsf{c},\mathsf{a}^{(k)})$ by

$$\lambda_2\,\mathfrak{B}(\mathsf{c})\rho'(1)\,\mathfrak{B}\,(\mathsf{a}^{(k)}) + \bar{\lambda}_2\,\mathfrak{B}(\mathsf{c})(\rho'(1)\,\mathfrak{B}\,(\mathsf{a}^{(k)}))^2 + e^{\frac{-K}{2}}$$
 Let $\xi>0$ satisfy

$$\eta := \lambda_2 \mathfrak{B}(\mathsf{c}) \rho'(1) + \bar{\lambda}_2 \mathfrak{B}(\mathsf{c}) (\rho'(1))^2 \xi < 1$$

and assume $\xi \geq \mathfrak{B}\left(\mathsf{a}^{(k)}\right)$ and K is large enough so that $\xi \geq \frac{2-\eta}{1-\eta}e^{-K/2}$. Then there exists N so that for $n \geq N$ we have $\mathfrak{B}\left(\mathsf{a}^{(n)}\right) \leq \frac{1}{1-\eta}e^{-K/2}$.

Minimum variable node degree equal to 3: Let us now assume that the minimum variable node degree is 3. Following the previous notation, we upper bound $\mathfrak{B}(a^{(k+1)})$ by

$$\lambda_3\,\mathfrak{B}(\mathsf{c})(\rho'(1)\,\mathfrak{B}\,(\mathsf{a}^{(k)}))^2 + \bar{\lambda}_3\,\mathfrak{B}(\mathsf{c})(\rho'(1)\,\mathfrak{B}\,(\mathsf{a}^{(k)}))^3 + e^{\frac{-K}{2}}$$

Let ξ be the positive solution to $\lambda_3 \mathfrak{B}(\mathsf{c}) \rho'(1)^2 \xi + \bar{\lambda}_3 \mathfrak{B}(\mathsf{c}) \rho'(1)^3 \xi^2 = 1/2$. Assume K large enough so that $2e^{-K/2} < \xi$. A little algebra shows that if $\mathfrak{B}(\mathsf{a}^{(0)}) \leq \xi$, then there exists N so that for all n > N we have

$$\mathfrak{B}(\mathsf{a}^{(n)}) \le 2e^{-K/2}, \qquad \mathfrak{B}(\mathsf{b}^{(n)}) \le 2\rho'(1)e^{-K/2}$$
 (3)

where the second inequality follows from the first and the additive bound on Bhattacharrya at check node.

This analysis can not show convergence to zero error although it can be used to show convergence to relatively small error rate. This is true even in the presence of degree two variable nodes, where zero error stability is not possible. For degree three and higher stability can be shown, but a refined analysis is needed.

C. Stability with Minimum Variable Degree Equal to Three

In this section we consider irregular ensembles where the minimum variable node degree is at least three. We generalize the standard stability analysis by separating out the saturated probability mass and tracking it through the variable node and check node updates. For simplicity we shall restrict to right regular ensembles. If the check node degree is d_r then K' is the magnitude of an outgoing message all of whose incoming messages have magnitude K. Although we focus on BP-like decoding our analysis applies to other algorithms such as minsum, in which case we have K' = K. In general, if $K_1, ..., K_d$ are the magnitudes of incoming messages then we assume that the outgoing magnitude satisfies $K' \leq \min_i \{K_i\}$ and $e^{-K'} \leq \sum_{i} e^{-K_i}$. Both conditions are satisfied by BP and min-sum. Messages entering a check node update a have the form $a = \gamma D(p, K) + \bar{\gamma} m$, where m is supported on (-K, K)and has total mass 1. Messages entering a variable node update b have the form $b = \gamma D(p, K') + \bar{\gamma} m$, where K' < K is the outgoing magnitude at a check when all incoming magnitudes equal K and m is supported on (-K', K'). From above, we have $e^{-K'} \le (d_r - 1)e^{-K}$ so that $K' \ge K - \ln(d_r - 1)$. Furthermore, we choose K large enough so that 2K' > K.

In the sequel we assume that the support of the channel c is (-K'', K''). The condition on K'' is that $K'' \leq 2K' - K$. The proof of the two statements below can be found in the upcoming paper [17].

1) Variable Node Analysis: Let us assume a variable node of degree d+1 and incoming density $\mathsf{b} = \gamma D(p,K') + \bar{\gamma}\mathsf{m}$. The outgoing density from the variable node has the form $\mathsf{a} = \gamma' D(p',K) + \bar{\gamma}'\mathsf{m}'$. Then we have

$$\begin{split} \bar{\gamma}' \, \mathfrak{B}(\mathsf{m}') & \leq \mathfrak{B} \, (\mathsf{c}) de^{-(d-1)(K/2 - \ln c)} (\bar{\gamma} \, \mathfrak{B}(\mathsf{m})) \\ & + (\gamma p) e^{-(d/2 - 1)(K - \ln c)} e^{d/2 \ln (3e)} (1 + 2 d_r), \\ \gamma' p' & \leq e^{-K/2} \, \mathfrak{B} \, (\mathsf{c}) de^{-(d-1)(K/2 - \ln c)} (\bar{\gamma} \, \mathfrak{B}(\mathsf{m})) \\ & + 2^d (\gamma p)^{\lfloor (d+2)/2 \rfloor} \, . \end{split}$$

- 2) Check Node Analysis: Let us assume a right regular ensemble with check degree d+1. Let us represent the density entering the check node as $\gamma D(p,K) + \bar{\gamma} m$ where m is a density supported on (-K,K). Then the density emerging out of the check node is given by $\gamma' D(p',K_d) + \bar{\gamma}' m' \triangleq (\gamma D(p,K) + \bar{\gamma} m)^{\boxplus d}$, where K_d is the magnitude, which satisfies $K \ln d \leq K_d \leq K$ and support of m is also $(-K_d,K_d)$. Then we have, $\bar{\gamma}' \mathfrak{B}(m') \leq \bar{\gamma} \mathfrak{B}(m) (\gamma C d^3 + d)$. Also, we note that $p' = p_d \leq dp$.
- 3) Proof of Stability for Degree Three+: Let us assume that the minimum variable node degree is at least three and a right regular degree d_r . In view of (3) we may assume $\mathfrak{B}\left(\mathsf{a}^{(n)}\right) \leq 2e^{-K/2}$ which implies $\mathfrak{B}\left(\mathsf{b}^{(n)}\right) \leq 2(d_r-1)e^{-K/2}$ for all $n \geq 0$. We write $\mathsf{a}^{(n)} = \gamma D(p,K) + \bar{\gamma}\mathsf{m}$ and now have the bound (for all n) $\gamma pe^K \leq 2$. Set $c = 2(d_r-1)$ and from the check node analysis above and the bound $\gamma pe^K \leq 2$ we get $\gamma C \leq 2(d_r-1)+1+\sqrt{d_r-1}$. Abusing the notation, set $C = 2(d_r-1)+1+\sqrt{d_r-1}$. Then $\mathsf{b}^{(n)} = \gamma' D(p',K_{d_r-1}) + \bar{\gamma}'\mathsf{m}'$ denote the density emerging from the check node update where the support of m' is $(-K_{d_r-1},K_{d_r-1})$. Then from the check node analysis above we get

$$\gamma' p' \le (d_r - 1)\gamma p$$

$$\bar{\gamma} \mathfrak{B} (\mathsf{m}') \le \bar{\gamma} \mathfrak{B} (\mathsf{m}) (\gamma C (d_r - 1)^3 + d_r - 1).$$

We choose a,b,c,d>0 arbitratily small. Then for all K large enough and for all $d\geq 2$ and d less then the maximum variable node degree, we have

$$(d_r - 1)2^d (2e^{-K/2})^{\lfloor d/2 \rfloor} \le a$$

$$\mathfrak{B}(\mathsf{c}) de^{-(d-1)(K/2 - \ln c)} (d_r - 1)C \le b$$

$$e^{-(\frac{d}{2} - 1)(K - \ln c)} e^{\frac{d}{2} \ln(3e)} (1 + 2d_r) 2(d_r - 1) e^{-K/2} \le c$$

$$\mathfrak{B}(\mathsf{c}) de^{-(d-1)(K/2 - \ln c)} (d_r - 1)C \le d$$

We now obtain for $n \geq 0$,

$$\begin{bmatrix} e^{K/2} \gamma p \\ \bar{\gamma} \, \mathfrak{B} \, (\mathsf{m}) \end{bmatrix}^{(n+1)} \leq \begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} e^{K/2} \gamma p \\ \bar{\gamma} \, \mathfrak{B} \, (\mathsf{m}) \end{bmatrix}^{(n)}$$

which is a stable system for suitable a, b, c, d. It is now easy to see that the bit error rate goes to zero since the probability mass on [-K, K) goes to zero as n becomes large.

VI. DISCUSSION

Consider a channel family, BMS(h), ordered by entropy and let h^{BP} denote the BP threshold when transmitting over this channel family using a (λ, ρ) LDPC ensemble such that the

variable degree is at least three. Note that when we consider a channel from BMS(h), we first (symmetric) saturate the channel to an appropriate finite support. One can show that as the support goes to $(-\infty, \infty)$, the saturated channel BP threshold goes to h^{BP} (cf. [17]). The above analysis shows that, for K large enough, the SatBP is successful for any channel with entropy arbitrarily close to channel corresponding to h^{BP} .

An interesting open question is to understand the scaling of K with the gap to capacity δ . From Lemma 11 we see that one would require the scaling of number of iterations wrt δ for unsaturated BP which is conjectured to scale like $O(1/\delta)$.

VII. ACKNOWLEDGEMENTS

SK would like to thank R. Urbanke for useful discussions.

REFERENCES

- T. Richardson and R. Urbanke, Modern Coding Theory. Cambridge University Press, 2008.
- [2] X. Zhang and P. Siegel, "Will the real error floor please stand up?" in Signal Processing and Communications (SPCOM), 2012 International Conference on, 2012, pp. 1–5.
- [3] B. Butler and P. Siegel, "Error floor approximation for ldpc codes in the awgn channel," in Communication, Control, and Computing (Allerton), 2011 49th Annual Allerton Conference on, 2011, pp. 204–211.
- [4] C. Schlegel and S. Zhang, "On the dynamics of the error floor behavior in (regular) ldpc codes," *Information Theory, IEEE Transactions on*, vol. 56, no. 7, pp. 3248–3264, 2010.
- [5] S. Zhang and C. Schlegel, "Controlling the error floor in ldpc decoding," Communications, IEEE Transactions on, vol. 61, no. 9, pp. 3566–3575, 2013.
- [6] X. Zhang and P. Siegel, "Quantized min-sum decoders with low error floor for ldpc codes," in *Information Theory Proceedings (ISIT)*, 2012 IEEE International Symposium on, 2012, pp. 2871–2875.
- [7] —, "Quantized iterative message passing decoders with low error floor for ldpc codes," pp. 1–14, 2013.
- [8] B. Vasic, D. V. Nguyen, and S. K. Chilappagari, "Failures and errorfloors of iterative decoders," Channel Coding: Theory, Algorithms, and Applications, Academic Press Library in Mobile and Wireless, Communications, Elsevier, New York, 2014.
- [9] J. Wang, T. Courtade, H. Shankar, and R. Wesel, "Soft information for ldpc decoding in flash: Mutual-information optimized quantization," in *Global Telecommunications Conference (GLOBECOM 2011)*, 2011 IEEE, 2011, pp. 1–6.
- [10] N. Kanistras, I. Tsatsaragkos, I. Paraskevakos, A. Mahdi, and V. Paliouras, "Impact of Ilr saturation and quantization on Idpc min-sum decoders," in *Signal Processing Systems (SIPS)*, 2010 IEEE Workshop on, 2010, pp. 410–415.
- [11] N. Kanistras, I. Tsatsaragkos, and V. Paliouras, "Propagation of Ilr saturation and quantization error in ldpc min-sum iterative decoding," in Signal Processing Systems (SiPS), 2012 IEEE Workshop on, 2012, pp. 276–281.
- [12] Y. Wu, L. Davis, and R. Calderbank, "On the capacity of the discretetime channel with uniform output quantization," in *Information Theory*, 2009. ISIT 2009. IEEE International Symposium on, 2009, pp. 2194– 2198.
- [13] T. Richardson and R. Urbanke, "The capacity of low-density parity check codes under message-passing decoding," *IEEE Trans. Inform. Theory*, vol. 47, no. 2, pp. 599–618, Feb. 2001.
- [14] S. Kudekar, T. Richardson, and R. L. Urbanke, "Spatially coupled ensembles universally achieve capacity under belief propagation," *CoRR*, vol. abs/1201.2999, 2012.
- [15] C. Villani, Optimal transport, Old and New. Springer, 2009, vol. 338.
- [16] K. Bhattad, V. Rathi, and R. Urbanke, "Degree optimization and stability condition for the min-sum decoderl," in *Proc. of the IEEE Inform. Theory Workshop*, 2007, conference, pp. 190–195.
- [17] S. Kudekar, T. Richardson, and A. Iyengar, "The effect of saturation on belief propagation decoding of ldpc codes," 2014, e-print: http://arxiv.org/.