# Minimization of the Worst Case Average Energy Consumption in UAV-Assisted IoT Networks

Osmel Martínez Rosabal<sup>®</sup>, Onel Luis Alcaraz López<sup>®</sup>, *Member, IEEE*, Dian Echevarría Pérez<sup>®</sup>, *Graduate Student Member, IEEE*, Mohammad Shehab<sup>®</sup>, Henrique Hilleshein, and Hirley Alves<sup>®</sup>, *Member, IEEE* 

Abstract—The Internet of Things (IoT) brings connectivity to a massive number of devices that demand energy-efficient solutions to deal with limited battery capacities, uplink-dominant traffic, and channel impairments. In this work, we explore the use of unmanned aerial vehicles (UAVs) equipped with configurable antennas as a flexible solution for serving lowpower IoT networks. We formulate an optimization problem to set the position and antenna beamwidth of the UAV, and the transmit power of the IoT devices subject to average-signal-toaverage-interference-plus-noise ratio (SINR) Quality-of-Service (QoS) constraints. We minimize the worst case average energy consumption of the latter, thus targeting the fairest allocation of the energy resources. The problem is nonconvex and highly nonlinear; therefore, we reformulate it as a series of three geometric programs that can be solved iteratively. Results reveal the benefits of planning the network compared to a random deployment in terms of reducing the worst case average energy consumption. Furthermore, we show that the target SINR is limited by the number of IoT devices, and highlight the dominant impact of the UAV hovering height when serving wider areas. Our proposed algorithm outperforms other optimization benchmarks in terms of minimizing the average energy consumption at the most energy-demanding IoT device, and convergence time.

Index Terms—Energy efficiency, geometric programming (GP), Internet of Things (IoT), reconfigurable antennas, unmanned aerial vehicle (UAV), worst case average energy consumption.

## I. INTRODUCTION

THE FIFTH generation of cellular networks (5G) is introducing for the first time, in addition to the traditional human-centric broadband communication services, new service classes related to the Internet of Things (IoT) [1]. IoT use cases are usually characterized by the deployment of numerous low-cost low-power devices, for which novel energy-efficient strategies are increasingly needed as the network densifies [2]–[4]. Furthermore, the information and communication technology industry currently contributes to 6% of global CO<sub>2</sub> emissions [5]. As a consequence, energy-efficient technologies and solutions are relentlessly pursued by industry and academy. We need to consider myriad of different approaches

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The authors are with the Centre for Wireless Communications, University of Oulu, 90570 Oulu, Finland (e-mail: osmel.martinezrosabal@oulu.fi; onel.alcarazlopez@oulu.fi; dian.echevarriaperez@oulu.fi; mohammad. shehab@oulu.fi; henrique.hilleshein@oulu.fi; hirley.alves@oulu.fi).

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to obtain environment-friendly IoT deployments including, but not limited to, reducing energy consumption, greener materials in the production of IoT devices, proper waste disposal, and sharing infrastructure [6], [7]. Focusing on the energy consumption component, IoT devices should just transmit necessary data while using efficient wake up protocols, sleep scheduling, collision/congestion-avoidance schemes, and other possible energy saving improvements [6].

The IoT traffic is uplink-dominant [8]; thus, significant energy saving is attained from reducing the IoT devices' transmit energy<sup>1</sup> [9]. Besides, with a network-wide reduced transmit energy, interference is reduced, which may help to sustain the desired Quality of Service (QoS) with fewer resources. Therefore, it is important to reduce the uplink transmit energy to reach more energy-efficient IoT solutions. However, depending on the distance and position of the IoT device with respect to its associated terrestrial base station (BS), this may be extremely difficult to achieve due to shadowing and blockage effects. The uplink channel may require to be compensated with greater transmit power [10]. In that sense, it is desirable to have a BS that can dynamically change its coverage based on the position and traffic pattern of the IoT devices. This goal can be achieved by using aerial BSs, such as unmanned aerial vehicles (UAVs) [11].

To this end, some authors recognize UAVs as a promising technology due to their potential to provide provisional communication infrastructure in disaster scenarios, as opportunistic relays to serve blocked links, or as flying BSs to boost coverage in certain areas [12]. Considering the foreseen massive number of IoT devices, UAVs are an attractive solution for energy efficiency and QoS improvements due to the enhanced coverage resulting from their high mobility and ability to hover as discussed in [13]. In fact, it is challenging to obtain line of sight (LOS) using terrestrial BSs in urban canyon environments [14], [15], and it is hard to envision smart cities without the assistance of UAVs [13], [16]. Furthermore, by equipping the UAVs with reconfigurable antennas [17], more degrees of freedom could be attained since it is possible to adjust the beam footprint of the UAV by means of electrical, optical, mechanical, and material change techniques to boost

<sup>&</sup>lt;sup>1</sup>Note that herein we refer to transmit energy to highlight that both, the transmit power and transmission duration, influence the energy consumption, and consequently the lifetime of the IoT device battery.

even more the coverage with QoS guarantees.<sup>2</sup> These features are precisely exploited in our work to reduce the worst case average energy consumption of the IoT devices.

#### A. Related Literature

Many recent works have addressed problems, such as the optimal UAVs' positioning and trajectory, and device association for reliable and energy-efficient communications. For instance, the work in [19] maximized the energy efficiency of the IoT network by associating UAVs and IoT devices using regret-matching learning. Al-Hourani et al. [20] derived the optimal hovering height of low-altitude UAVs to achieve the maximum coverage based on the path loss, elevation angle and statistical parameters of the urban environment. In [21], an approach for maximizing the coverage area was derived based on the optimal deployment of multiple UAVs at a fixed altitude with directional antennas. An energy-efficient 3-D placement study was carried out in [22] to maximize the number of covered users with a minimum transmit power. Meanwhile, UAVs are shown in [23] to save valuable energy resources to ground devices in hostile or inaccessible places. Authors discussed the tradeoff between system efficiency and energy efficiency, considering the UAV altitude and speed, and also the frame length at the MAC layer. Alsamhi et al. [24] used artificial neural networks (ANNs) to predict the coverage and the received signal strength of IoT devices based on the altitude and distance from the UAV. They demonstrated that ANNs can predict better the path loss of the link between the IoT devices and the UAV than the empirical Hata model. Moreover, a better UAV altitude prediction leads to satisfy the OoS requirements more easily. Also, the authors estimated the probability of LOS for suburban, urban, and dense urban environments in function of the elevation angle, and showed that given an elevation angle closer to 90°, the probability of LOS tends to be 1 for any of the studied urban environments. Alzenad et al. [25] compared the performance of exhaustive search (ES) and maximal weighted area (MWA) in finding the optimal altitude for UAVs to provide coverage to devices with different QoS requirements. Their numerical simulations showed that MWA algorithm has a close performance to ES algorithm while it has lower complexity. Sikeridis et al. [26] proposed a reinforcement-learning (RL)-based approach to jointly optimize the IoT nodes' transmission power and the UAV positioning. Sun et al. [27] studied the joint age of information and IoT devices' energy consumption minimization in a UAV-assisted data collection problem. Therein, the authors trained an ANN to find the optimal trajectory of the UAV as well as the resource (bandwidth and transmit power) allocation for data collection. An energy-efficient UAV-enabled solution for massive IoT-shared spectrum access was proposed in [28], where the IoT devices' transmit power was optimized while the interference constraint for the closest primary user is respected.

<sup>2</sup>We acknowledge that some design challenges impact the performance of reconfigurable antennas, such as the proper design of a biasing network, difficult integration, high power consumption, and mechanical stress of moving parts. However, recent works have shown how to ease the design process of reconfigurable antennas. See [18] and references therein for more information.

Moreover, UAVs may be used as data aggregators. For instance. Wang et al. [29] proposed an energy-efficient solution for IoT data collection at cell edges. Therein, the trajectory of the UAV and the IoT devices transmission schedule is optimized. In [30], UAVs were used as communication relays to assist the links between smart devices and a low-orbiting satellite in scenarios without coverage from terrestrial BSs. The authors optimized the subchannel selection, UAV relays deployment and uplink transmission power control of smart devices to maximize the energy efficiency of the system. Azizi et al. [31] minimized the overall transmit power of IoT devices considering the radio resource allocation, 3-D placement and user association to the UAV BS. They used semidefinite relaxation and geometric programming (GP) to solve the corresponding optimization problem. However, the works in [19]-[26], [28]-[31] do not consider the activation pattern of the IoT devices, which considerably influences the optimum system setup. Conversely, Mozaffari et al. [32] iointly determined the optimal UAV's location, device association and uplink power control considering the activation patterns of IoT devices and a channel assignment strategy. They transformed this nonconvex optimization problem into convex by decomposing it in two subproblems. First, they considered that the UAV BSs are in a fixed position to find the jointly optimal device associations and devices' transmit power. Second, the positioning of the UAVs was optimized given the device associations. However, they minimized the total transmission power, which could cause unfairness among IoT nodes because UAV BSs would focus on big clusters of IoT nodes to reduce the overall transmission power.

Recently, Wang et al. [33] considered the use of UAVs as providers of computational resources for terrestrial devices that can either upload their tasks to the UAVs or compute them locally. Therein, they jointly minimize the total energy consumption spent for uplink transmissions at the served devices and the trajectory of the UAVs using a deep RL approach. Finally, He et al. [34] studied a multiuser network served by a UAV which is equipped with a reconfigurable antenna. Therein, the authors jointly optimize the UAV's hovering height and antenna beamwidth to maximize the throughput under different multiuser communication models. They divided the network into nonoverlapping clusters and then proposed a fly-hover-and-communicate protocol for the UAV to sequentially serve each cluster.

## B. Contributions

Different from the above works, herein, we propose a fair energy-efficient UAV-assisted IoT network, where we reduce the transmit energy consumption for the worst case IoT nodes. In this case, the UAV BS is repositioned considering the worst case average energy consumption of the IoT nodes' uplink transmission and their activation patterns. In order to achieve this, we consider a UAV BS equipped with a reconfigurable antenna and serving multiple IoT devices. This is motivated by the increasing interest on incorporating reconfigurable antennas at UAVs [35] and/or creating antenna arrays via UAV swarms [36]. By using a reconfigurable antenna, the

UAV can properly vary the beamwidth to optimize the system performance. In this case, the optimization is in terms of the energy efficiency measured as the average energy consumption at the most energy-demanding IoT device when transmitting in the uplink. The optimal UAV 3-D position and devices' transmit power are found based on their activation pattern and subject to average-signal-to-average-interference-plus-noise ratio  $(\bar{S}\bar{IN}R)$  OoS constraints.

Our main contributions are fourfold.

- Instead of the traditional total power minimization problem as in [32], we aim at reducing the average energy consumption at the most energy-demanding device in the network. This conduces to fair allocation of the power resources, which allows synchronizing the devices' lifetime so that maintenance (e.g., for battery replacement) can be efficiently planned.
- The resulting optimization problem, which is not convex and highly nonlinear, is approximately recast as a series of three GPs that can be efficiently solved.
- 3) The proposed algorithm reaches near-global optimal solutions for the worst case average energy consumption, and considerably outperforms other benchmark schemes based on interior-point methods (IPMs) and genetic algorithms (GAs) in terms of minimizing the worst case average energy consumption of the IoT devices and the computation time.
- 4) Results show that the number of IoT devices limits the achievable QoS due to interference, and that the worst case average energy consumption does not depend significantly on the density of the obstacles but on their height. The optimal hovering height of the UAV increases linearly with respect to the coverage area and, hence, the worst case average energy consumption of the uplink transmissions grows with the distance between the UAV and the IoT devices.

#### C. Outline

The remainder of this article is structured as follows. Section II describes the system model and presents the problem formulation. Section III discusses some insights on the problem feasibility conditions and reformulates the problem as a series of GPs. Section IV presents the proposed optimization algorithm, while simulation and numerical results are analyzed in Section V. Finally, we draw conclusions and make final remarks in Section VI. To make this article more tractable, we summarize the key abbreviations and symbols that will appear throughout this article in Table I.

## II. SYSTEM LAYOUT AND PROBLEM FORMULATION

#### A. System Layout

We consider a wireless system consisting of a set  $\mathcal{K} = \{1, 2, ..., K\}$  of K low-power single-antenna IoT nodes, whose deployment in the 2-D plane is given by  $\{(x_k, y_k) | k \in \mathcal{K}\}$ . An IoT node is considered to be active when it has data to transmit [37]. Note that not all devices are active at the same time; thus, hereafter,  $c_k \in (0, 1)$  denotes the probability of device

TABLE I
IMPORTANT ABBREVIATIONS AND SYMBOLS

GA	Genetic Algorithms
GP	Geometric Programming
IoT	Internet of Things
IPMs	Interior-point methods
LAP	Low-altitude platform
LOS	Line of Sight
QoS	Quality of Service
$\overline{SINR}$	average-Signal-to-average-Interference-plus-Noise Ratio
UAV	Unmanned Aerial Vehicle
inf	Infimum
sup	Supremum
$c_k$	Activation probability of device $k$
$G_k$	Antenna gain seen by device k
$L_k$	Path loss between the UAV and device $k$
$\mathcal{O}\left(\cdot\right)$	Order of the function
$p_k$	Transmit power of device $k$
$eta, \psi$	Propagation parameters
$\theta_B$	UAV's antenna half beamwidth
$\eta$	Path loss coefficient
ζ	Convergence parameter
K	Number of IoT devices
$\gamma_k$	$\overline{\text{SINR}}$ of device $k$

 $k \in \mathcal{K}$  being active. This information could be acquired beforehand by applying a traffic prediction algorithm that depends on correlated devices activity or prior knowledge of devices activation patterns as in [37]–[39]. The activation pattern is herein exploited to efficiently allocate resources to IoT nodes [37]. The same applies for downlink communications in temporally crowded places due to major events (e.g., sport matches and concerts) where UAVs could be sent to offload the existing permanent wireless network [40]. In both uplink and downlink cases for UAV-assisted wireless network, we can use the data traffic prediction to efficiently allocate resources and to reposition the UAVs.

We analyze an uplink scenario as illustrated in Fig. 1, where active devices communicate over the same frequency band<sup>3</sup> with a rotary-wing UAV at height<sup>4</sup> h. We assume the use of a low-altitude platform (LAP) such as a quadrotor UAV. The position of the UAV is then fully given by  $(x_{uav}, y_{uav}, h)$ . Additionally, we denote the UAV's directional antenna half beamwidth by  $\theta_B$ ; thus, the antenna gain seen by the kth IoT node transmissions can be approximated by

$$G_k = \begin{cases} G_{3dB}, & -\frac{\theta_B}{2} \le \varphi_k \le \frac{\theta_B}{2} \\ 0, & \text{otherwise} \end{cases}$$
 (1)

<sup>&</sup>lt;sup>3</sup>The extension to a multichannel scenario with random or deterministic channel allocation is straightforward.

<sup>&</sup>lt;sup>4</sup>The term "height" refers to the vertical distance from the surface where the IoT nodes are deployed to the UAV. However, the regional operational rules for flying small UAVs define, in most cases, the hovering height limits in terms of the absolute altitude, which is the height of the UAV above sea level. See [41], [42] for more information.

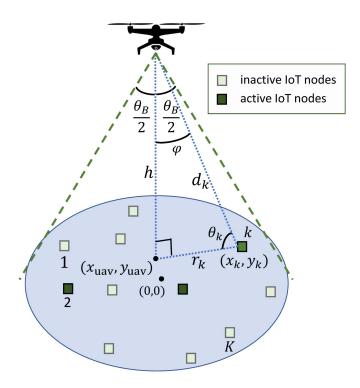


Fig. 1. System model comprises a set  $\mathcal K$  of IoT nodes served by a rotary-wing UAV.

where  $\varphi_k$  is the corresponding sector angle.  $G_{3\text{dB}}$  is the main lobe gain, which we consider as a null gain outside the main lobe, and is given approximately by  $G_{3\text{dB}} \approx (8.83/\theta_B^2)$  with  $\theta_B$  in radians [21], [43]. Additionally, the UAV is equipped with a reconfigurable antenna such that it is capable of tuning  $\theta_B \geq \theta_0$  as it sees fit, where  $\theta_0$  is the minimum antenna half beamwidth. The reader can refer to [44] and the references therein for more information about different techniques for implementing reconfigurable antennas.

## B. Channel Model

The ground-to-air channel depends greatly on the type of environment (e.g., rural, suburban, urban, high-rise urban, etc). Note that in such practical scenarios one may not have any additional information about the exact locations, heights, and number of the obstacles. Therefore, it is advisable to consider the randomness associated with the LOS and non-LOS (NLOS) links when designing the UAV-based communication system.

For ground-to-air communications, each device will typically have a LOS view towards the UAV with a given probability. This LOS probability depends not only on the environment but also on the elevation angle, and for the *k*th IoT device it is commonly modeled as [32]

$$P_k^{\text{los}} = \frac{1}{1 + \psi e^{-\beta(\theta_k - \psi)}} \tag{2}$$

where  $\psi$  and  $\beta$  are constant values, which depend on the carrier frequency and type of environment, while

$$\theta_k = \tan^{-1} \frac{h}{r_k} = \sin^{-1} \frac{h}{d_k} \tag{3}$$

is the elevation angle. Note that

$$d_k = \sqrt{r_k^2 + h^2} \tag{4}$$

$$r_k = \sqrt{(x_k - x_{\text{uav}})^2 + (y_k - y_{\text{uav}})^2}$$
 (5)

denote the distance from  $k \in \mathcal{K}$  to the UAV and to its projection on ground, respectively. Then, the NLOS probability is given by  $P_k^{\text{nlos}} = 1 - P_k^{\text{los}}$ . As expected, the LOS probability in (2) models practical phenomena since by increasing the elevation angle and/or the UAV altitude, the chances of LOS are greater.

The path-loss model for LOS and NLOS links between device k and the UAV is given by [45]

$$L_k = \eta \left(\frac{4\pi f_c d_k}{c}\right)^{\alpha} \tag{6}$$

where  $f_c$  is the carrier frequency,  $\alpha$  is the path-loss exponent, c is the speed of light, while  $\eta \in \{\eta_1, \eta_2\}$ , where  $\eta_1$  and  $\eta_2$  ( $\eta_2 > \eta_1 > 1$ ) are the excessive path-loss coefficients under LOS and NLOS conditions, respectively. Now, leveraging (2) and (6), the average path loss between device k and the UAV can be expressed as

$$\bar{L}_k = \left(P_k^{\log} \eta_1 + P_k^{\log} \eta_2\right) (\kappa d_k)^{\alpha} \tag{7}$$

where  $\kappa = 4\pi f_c/c$ . Then, the average channel power gain is given by  $\bar{g}_k = 1/\bar{L}_k$ . Finally, the per-link communication performance is measured through its  $\bar{S}\overline{IN}R$ , which for the kth IoT device, it is given by

$$\gamma_k = \frac{G_k p_k \bar{g}_k}{\sum_{j \in \mathcal{K} \setminus k} c_j G_j p_j \bar{g}_j + \sigma^2}$$
 (8)

where  $p_k$  denotes its transmit power and  $\sigma^2$  is the additive white Gaussian noise power at the UAV receiver.

#### C. Problem Formulation

Herein, we are interested in minimizing the worst case average energy consumption per device by optimizing not only their transmit power but also the UAV position  $x_{uav}$ ,  $y_{uav}$ , h and antenna beamwidth  $\theta_B$ . To this end, we cast the min–max optimization problem as follows:

**P1:** minimize 
$$\max_{x_{\text{uav}}, y_{\text{uav}}, h, \{p_k\}, \theta_B} \{c_k p_k\}$$
 (9a)

subject to 
$$\gamma_i \ge \gamma_0 \quad \forall i \in \mathcal{K}$$
 (9b)

$$p_{\min} \le p_i \le p_{\max} \quad \forall i \in \mathcal{K}$$
 (9c)

$$h \ge h_{\min}$$
 (9d)

$$\theta_B > \theta_0.$$
 (9e)

Herein, we consider that the transmission time is normalized, which allows us to use  $\max_{k \in \mathcal{K}} \{c_k p_k\}$  as the average energy consumption of the most energy-demanding device

 ${}^5\mathrm{The}$  term  $\bar{\mathrm{SINR}}$  here is used to highlight the fact that instead of using  $L_k$  separately for LOS and NLOS links we utilize  $\bar{L}_k$ , while accounting also for the average activation probabilities. By doing this, the SINR expression becomes more tractable than the *average* SINR which is more naturally linked to decoding success. Hence,  $\gamma_0$  is chosen such that when  $\bar{\mathrm{SINR}} \geq \gamma_0$ , the chances of outage are negligible. Such an approach has been also adopted in several works, e.g., [20], [22], [32], [45].

of the network. By minimizing this quantity, we limit the average energy consumption at each device which ultimately prevent long-lasting peaks in their energy consumption profile that quickly reduce their batteries' lifetime [46]. Besides, we assume that the IoT devices contend for the same uplink resource block using a grant-free random access protocol, whose recovery performance at the UAV depends on  $\gamma_0$ in (9b) [47], i.e., an  $\overline{SINR}$  target that must be achieved by all devices.<sup>6</sup> Power restrictions in (9c) are due to hardware limitations and/or spectrum regulations, while the UAV altitude restriction in (9d) is due to aviation regulations [10], and the antenna beamwidth constraint in (9e) is inherent to the antenna hardware. Finally, we assume that the system dynamics are quasistatic.

Since neither the objective function (9a) nor the inequality constraints (9b) are convex, P1 is obviously not convex. This nonconvexity and the extreme nonlinearity of (9b) on all the optimization variables, render P1 extremely hard to solve efficiently in its current form. Therefore, in the following section, we aim at solving such issue by approximately casting P1 as a series of GP problems, which, in turn, can be efficiently solved.

## III. PROBLEM REFORMULATION

## A. Insights on Problem Feasibility

We start by studying the feasibility of the problem. We observe that  $\gamma_k$  in (8) is a step function of  $G_{3dB}$  since we assume null gain outside the main antenna lobe. Then, for guaranteeing (9b), all the IoT devices must be in the ground footprint of the UAV's antenna main lobe. Such a practical constraint can be geometrically given as

$$\theta_B \ge 2 \tan^{-1} \frac{\max_k r_k}{h} \tag{10}$$

and then, we are free to modify (8) as follows:

$$\gamma_k = \frac{G_{3\text{dB}}p_k\bar{g}_k}{G_{3\text{dB}}\sum_{j\in\mathcal{K}\backslash k}c_jp_j\bar{g}_j + \sigma^2}.$$
 (11)

The next proposition establishes the maximum target  $\overline{SINR}$ for any IoT node in our setup.

*Proposition 1:* For feasibility, the target  $\overline{SINR}$  is required to satisfy the following constraint in the considered scenario:

$$\gamma_0 \le \frac{1}{(K-1) \max_k c_k + \frac{\sigma^2 \theta_0^2}{8.83 p_{\text{max}}}}$$
(12)

$$<\frac{1}{(K-1)\max_k c_k}. (13)$$

Proof: Let us proceed as follows. From (11), we have

$$\gamma_0 \leq \inf_{k \in \mathcal{K}} \left\{ \sup_{p_k, c_k, g_k, \theta_B} \{ \gamma_k \} \right\}$$

<sup>6</sup>Herein, we consider a common SINR target constraint for all devices possibly corresponding to the same application—for solving the minimization of the worst case average energy consumption of the massive IoT deployment. However, we acknowledge that IoT networks have heterogeneous QoS requirements and, thus, we left this problem for future works.

$$\frac{(a)}{=} \sup_{p_{k}, c_{k}, g_{k}, \theta_{B}} \left\{ \frac{G_{3dB}p_{k} * c_{k} * \bar{g}_{k} * / c_{k} *}{G_{3dB}c_{k} * p_{k} * \sum_{j \in \mathcal{K} \setminus k} * \bar{g}_{j} + \sigma^{2}} \right\}$$

$$\frac{(b)}{=} \sup_{\theta_{B}} \left\{ \frac{1}{(K-1) \max_{k} c_{k} + \frac{\sigma^{2}}{G_{3dB}p_{\max}}} \right\}$$

$$\frac{(c)}{=} \frac{1}{(K-1) \max_{k} c_{k} + \frac{\sigma^{2}\theta_{0}^{2}}{8.83p_{\max}}} \tag{14}$$

where the inf operation exists because all nodes are required to satisfy the QoS constraint  $\gamma \geq \gamma_0$ . Then, (a) comes from assuming that  $k^* \in \mathcal{K}$  is the IoT node with the weakest  $\bar{g}_k$ and using the fact that in the best scenario the solution of P1 leads to  $c_k p_k = \mu \ \forall k \in \mathcal{K}$ . Next, (b) comes from assuming all nodes with equal path loss and maximum transmit power, which maximizes the expression given in (a). Then, we attain (c) after letting  $\theta_B \to \theta_0$ , which matches (12); while (13) is a relaxed result that allows for establishing a preliminary bound without the knowledge of  $\sigma^2$ ,  $\theta_0$ , and  $p_{\text{max}}$ .

Note that (13) becomes tight as K,  $\max_k c_k$ , and  $p_{\max}$ increase and/or  $\sigma^2$  and  $\theta_0$  decrease.

#### B. Geometric Program Formulation

First, let us assume that P1 can be partitioned into two GP sub-problems as follows.

1) **P1-1**, which is **P1** given  $x_{uav}$ ,  $y_{uav}$ . This requires initializing  $x_{uav}$ ,  $y_{uav}$ , which can be done by simply choosing it such that  $[x_j, y_j]^T \leq [x_{uav}, y_{uav}]^T \leq [x_k, y_k]^T$ , for certain  $j, k \in \mathcal{K}$ . Intuitively, the UAV's projection on ground is expected to be close to a centroid determined by all IoT nodes position. Therefore, and given the devices activation probability a good initialization for  $x_{uav}$ ,  $y_{uav}$ can be

$$x_{\text{uav}}^{(0)} = \frac{\sum_{k=1}^{K} c_k x_k}{\sum_{k=1}^{K} c_k}, \quad y_{\text{uav}}^{(0)} = \frac{\sum_{k=1}^{K} c_k y_k}{\sum_{k=1}^{K} c_k}.$$
 (15)

2) **P1-2**, which is **P1** with optimization variables  $x_k$ ,  $y_k$ , and  $p_k \ \forall k \in \mathcal{K}$ , where h and  $\theta_B$  are outputs of **P1-1**.

Note that if such a problem partition exists as will be shown later, then P1 can be solved by mutually projecting the subproblems' solutions in an iterative way. This is because each GP subproblem can be readily transformed to a convex problem (see Section IV), and then according to Newmann's alternating projection lemma, the resulting iterative procedure always converges to the global optimum [48].

1) **P1-1**: We depart from **P1** (9) by noting that the unconstrained optimization part in (9a) can be alternatively stated in the GP standard form as

minimize 
$$\max_{k \in \mathcal{K}} \{c_k p_k\} \to \text{minimize } t$$
 (16a)

subject to 
$$c_k p_k t^{-1} \le 1 \quad \forall k \in \mathcal{K}$$
 (16b)

while the constraints (9c)-(9e) transform to

$$p_{\min}p_i^{-1} \le 1 \quad \forall i \in \mathcal{K} \tag{17a}$$

$$p_i p_{\text{max}}^{-1} \le 1 \quad \forall i \in \mathcal{K} \tag{17b}$$

$$p_i p_{\text{max}}^{-1} \le 1 \quad \forall i \in \mathcal{K}$$
 (17b)  
 $h_{\text{min}} h^{-1} \le 1$  (17c)

$$\theta_0 \theta_R^{-1} \le 1. \tag{17d}$$

Afterward, we are required to deal just with (9b), which is very intricate. As commented in the previous section, such constraint can be divided into two constraints, which are given in (10) and (11). Although, the first one already involves a tangent function, which is not allowed in a GP environment, we can take advantage of the limited range of  $\theta_B$  values, e.g.,  $\theta_B \in (\theta_0, 2\tan^{-1}[(\max_{i \in \mathcal{K}} r_i)/h_{\min}]]$ , to use standard curve fitting tools and write  $\tan(\theta_B/2) \approx q_1\theta_B^{q_2}$ . Note that  $\theta_B < \pi$  since  $h_{\min} > 0$ , which favors the adopted power approximation. Also,  $q_1$  and  $q_2$  are positive since for a feasible  $\theta_B$  the function is increasing and positive. The approximation can be tight for  $\max_{i \in \mathcal{K}} r_i \leq 2h_{\min}$ , which should hold in practical setups. Therefore, (10) is relaxed to

$$q_1^{-1}h^{-1}\theta_B^{-q_2} \max_{i \in \mathcal{K}} r_i \le 1.$$
 (18)

Regarding the constraints related to  $\gamma_i \ge \gamma_0$ , where  $\gamma_i$  is given in (11), we proceed as follows:

$$\frac{G_{3dB}p_{k}\bar{g}_{k}}{G_{3dB}\sum_{j\in\mathcal{K}\setminus k}c_{j}p_{j}\bar{g}_{j}+\sigma^{2}} \geq \gamma_{0}$$

$$\gamma_{0}\sum_{j\in\mathcal{K}\setminus k}c_{j}p_{j}\bar{g}_{j}p_{k}^{-1}\bar{g}_{k}^{-1}+\gamma_{0}\sigma^{2}G_{3dB}^{-1}p_{k}^{-1}\bar{g}_{k}^{-1}\leq 1$$

$$\begin{cases}
\gamma_{0}\sum_{j\in\mathcal{K}\setminus k}c_{j}p_{j}\bar{g}_{j}p_{k}^{-1}u_{k}+\gamma_{0}\sigma^{2}G_{3dB}^{-1}p_{k}^{-1}u_{k} & \leq 1\\ \bar{g}_{k}^{-1}u_{k}^{-1} & \leq 1
\end{cases}$$
(19)

where the last transformation comes from introducing the auxiliary variables  $\{u_k\}$ . Then, now it is just a matter of expressing  $\bar{g}_k$  in a posynomial form [49], which we address as follows:

$$\bar{g}_{k} = \frac{1}{\bar{L}_{k}} = \frac{\kappa^{-\alpha} d_{k}^{-\alpha}}{P_{k}^{\log} \eta_{1} + P_{k}^{\log} \eta_{2}}$$

$$\stackrel{(a)}{=} \frac{\kappa^{-\alpha} d_{k}^{-\alpha} \left(1 + \psi e^{\psi \beta} e^{-\beta \theta_{k}}\right)}{\eta_{1} + \eta_{2} \psi e^{\psi \beta} e^{-\beta \theta_{k}}}$$

$$\stackrel{(b)}{\leq} (r_{k}h)^{-\alpha/2} \delta \tag{20}$$

where  $\delta = [(2^{-\alpha/2}\kappa^{-\alpha}(1+\psi e^{\psi\beta}))/(\eta_1+\eta_2\psi e^{\psi\beta})]$ . Note that (a) comes from using (7) and (2), while (b) follows after using the inequality between the arithmetic and geometric means:  $d_k^{-\alpha} = (r_k^2 + h^2)^{-\alpha/2} \leq (2r_k h)^{-\alpha/2}$ , and taking advantage of

$$\frac{1 + \psi e^{\psi \beta} e^{-\beta \theta_k}}{\eta_1 + \eta_2 \psi e^{\psi \beta} e^{-\beta \theta_k}} \le \frac{1 + \psi e^{\psi \beta}}{\eta_1 + \eta_2 \psi e^{\psi \beta}} = 2^{\alpha/2} \kappa^{\alpha} \delta.$$

Note that by using the upper bound of  $\bar{g}_k$  provided in (20), we still guarantee that all the constraints of the original problem are satisfied. However, since we constrained further the feasibility set, we may find a nonglobal optimum solution, which is the cost paid for our simplifications. Now, we can state **P1-1** as a GP as given next

**P1-1:** minimize 
$$h,\{p_k\},\theta_B,t,\{u_k\}$$
 (21a)

subject to 
$$c_i p_i t^{-1} \le 1 \quad \forall i \in \mathcal{K}$$
 (21b)

$$p_{\min}p_i^{-1} \le 1 \quad \forall i \in \mathcal{K} \tag{21c}$$

$$p_i p_{\text{max}}^{-1} \le 1 \quad \forall i \in \mathcal{K}$$
 (21d)

$$h_{\min}h^{-1} \le 1 \tag{21e}$$

$$\theta_0 \theta_R^{-1} \le 1 \tag{21f}$$

$$q_1^{-1}h^{-1}\theta_B^{-q_2}\max_{i\in\mathcal{K}}r_i \le 1$$
 (21g)

$$\omega(h, \{p_i\}, \theta_B, u_i) \le 1 \quad \forall i \in \mathcal{K} \quad (21h)$$

$$u_i^{-1} r_i^{\alpha/2} h^{\alpha/2} \delta^{-1} \le 1 \quad \forall i \in \mathcal{K}$$
 (21i)

where (21h) and (21i) come from substituting (20) into (19). Then, we have

$$\omega(h, \{p_i\}, \theta_B, u_i) = \gamma_0 h^{-\alpha/2} \delta p_i^{-1} u_i \sum_{j \in \mathcal{K} \setminus i} c_j p_j r_j^{-\alpha/2} + \gamma_0 \sigma^2 G_{3dB}^{-1} p_i^{-1} u_i.$$
 (22)

Finally, note that **P1-1** is a GP problem with 2K + 3 variables and 5K + 3 inequality constraints.

2) **P1-2**: Now, given the optimization results from **P1-1** and departing from **P1** (9), we formulate **P1-2** in order to find the optimum UAV position and power allocation profile. Note that similar to **P1-1**, herein (16) and (17) also hold. Meanwhile, without loss of generality we assume positive coordinates,  $x_k, y_k \ge 0 \ \forall k \in \mathcal{K}$  such that obviously  $x_{uav}, y_{uav} \ge 0$  holds as well. Then, we define  $\tilde{x}_k = |x_k - x_{uav}|$  and  $\tilde{y}_k = |y_k - y_{uav}|$  such that for given  $\theta_B$  and h and with the help of (5), constraint (10) can be rewritten as

$$\tan \frac{\theta_B}{2} \ge \frac{r_k}{h}$$

$$\left(h \tan \frac{\theta_B}{2}\right)^{-1} \sqrt{\tilde{x}_k^2 + \tilde{y}_k^2} \le 1$$

$$\left(h \tan \frac{\theta_B}{2}\right)^{-2} \tilde{x}_k^2 + \left(h \tan \frac{\theta_B}{2}\right)^{-2} \tilde{y}_k^2 \le 1$$
(23)

while instead of  $\tilde{x}_k = |x_k - x_{uav}|$ , we use

$$\tilde{x}_{k} = |x_{k} - x_{uav}| \to \begin{cases}
\tilde{x}_{k} \ge x_{k} - x_{uav} \\
\tilde{x}_{k} \ge x_{uav} - x_{k}
\end{cases}$$

$$\to \begin{cases}
\frac{\tilde{x}_{k}}{x_{k}} + \frac{x_{uav}}{x_{k}} \ge 1 \\
\frac{\tilde{x}_{k}}{x_{uav}} + \frac{x_{k}}{x_{uav}} \ge 1
\end{cases}$$

$$\overset{\sim}{\to} 8\begin{cases}
2x_{k}^{-1} \tilde{x}_{k}^{1/2} x_{uav}^{1/2} \ge 1 \\
2x_{uav}^{-1} \tilde{x}_{k}^{1/2} x_{uav}^{1/2} \ge 1
\end{cases}$$

$$\to \begin{cases}
\frac{1}{2} x_{k} \tilde{x}_{k}^{-1/2} x_{uav}^{-1/2} \le 1 \\
\frac{1}{2} x_{uav} \tilde{x}_{k}^{-1/2} x_{uav}^{-1/2} \le 1
\end{cases}$$
(24)

and the same applies for  $\tilde{y}_k = |y_k - y_{uav}|$ .

Regarding the constraints related with  $\gamma_i \ge \gamma_0$ , where  $\gamma_i$  is given in (11), note that (19) still holds, while (20) can be further transformed to

$$\bar{g}_k \le (2\tilde{x}_k \tilde{y}_k)^{-\alpha/4} h^{-\alpha/2} \delta \tag{25}$$

by using  $r_k^{-\alpha/2}=(\tilde{x}_k^2+\tilde{y}_k^2)^{-\alpha/4}\leq (2\tilde{x}_k\tilde{y}_k)^{-\alpha/4}$ . Then, we can state **P1-2** as a GP as given next

P1-2: 
$$\underset{\{x_{\text{uav}}, y_{\text{uav}}, \tilde{x}_k, \tilde{y}_k, p_k, u_k\}, t}{\text{subject to } c_i p_i t^{-1} \le 1}$$
 (26a)

<sup>&</sup>lt;sup>7</sup>It can be straightforwardly performed by properly setting the origin of coordinates.

<sup>&</sup>lt;sup>8</sup>Herein, we use the inequality between the arithmetic and geometric means. Note that this somewhat reduces the constraint set, but we deal with it later in Section IV.

# Algorithm 1 Optimum UAV Position and IoT Nodes' Transmit Power

- 1: **Input:**  $\{x_k, y_k, c_k\}_{\forall k \in \mathcal{K}}, \gamma_0, p_{\min}, p_{\max}, h_{\min}, \xi\}$
- 2: it = 0 (iteration index) 3: Set  $t^{(0)} = \infty$  and  $x_{\text{uav}}^{(0)}$ ,  $y_{\text{uav}}^{(0)}$  according to (15)
- 5: it  $\leftarrow$  it +1
- Solve **P1-1** given  $x_{\text{uav}}^{(\text{it}-1)}$ ,  $y_{\text{uav}}^{(\text{it}-1)}$ ,  $h^{(\text{it})}$ ,  $\{p_k^{(\text{it})}\}$ ,  $\theta_B^{(\text{it})}$ output:
- $h^{(k)}, \{p_k^{(k)}\}, \sigma_B$ Update:  $\theta_B^{(it)} \leftarrow \max\left(2\tan^{-1}\frac{\max_k r_k}{h^{(it)}}, \theta_0\right)$ 7:
- Solve **P1-2** given  $h^{(it)}$ ,  $\theta_B^{(it)}$ , output:

- 8: Solve F1-2 given n,  $v_B$ ,  $\{p_k^{(it)}\}$ ,  $t^{(it)}$ 9: **until**  $t^{(it-1)} t^{(it)} \le \xi$ 10: Solve **P1-3** given  $h^{(it)}$ ,  $\theta_B^{(it)}$ ,  $x_{uav}^{(it)}$ ,  $y_{uav}^{(it)}$ , **output:**  $\{p_k^*\}$ 11: **Output:**  $\{p_k^*\}$ ,  $x_{uav}^* = x_{uav}^{(it)}$ ,  $y_{uav}^* = y_{uav}^{(it)}$ ,  $\theta_B^* = \theta_B^{(it)}$ ,  $t = x_{uav}^*$

$$p_{\min} p_i^{-1} \le 1$$
 (26b)  
 $p_i p_{\max}^{-1} \le 1$  (26c)

$$p_i p_{max}^{-1} < 1$$
 (26c)

$$\left(h\tan\frac{\theta_B}{2}\right)^{-2}\tilde{x}_i^2 + \left(h\tan\frac{\theta_B}{2}\right)^{-2}\tilde{y}_i^2 \le 1$$

$$\frac{1}{2}x_k\tilde{x}_k^{-1/2}x_{\text{uav}}^{-1/2} \le 1 \tag{26e}$$

$$\frac{1}{2}x_k\tilde{x}_k^{-1/2}x_{\text{uav}}^{-1/2} \le 1$$
 (26e)  
$$\frac{1}{2}x_{\text{uav}}\tilde{x}_k^{-1/2}x_k^{-1/2} \le 1$$
 (26f)

$$\frac{1}{2} y_k \tilde{y}_k^{-1/2} y_{\text{uav}}^{-1/2} \le 1 \tag{26g}$$

$$\frac{1}{2}y_{\text{uav}}\tilde{y}_k^{-1/2}y_k^{-1/2} \le 1 \tag{26h}$$

$$\bar{\tilde{\omega}}(\tilde{x}_i, \tilde{y}_i, p_i, u_i) < 1 \tag{26i}$$

$$(2\tilde{x}_i \tilde{y}_i)^{\alpha/4} u_i^{-1} h^{\alpha/2} \delta^{-1} \le 1 \tag{26j}$$

where the constraints are  $\forall i \in \mathcal{K}$  and  $\tilde{\omega}$  is given by  $\omega$  in (22), but for fixed h,  $\theta_B$  and by substituting  $r_i^{-\alpha/2}$  by  $(2\tilde{x}_i\tilde{y}_i)^{-\alpha/4}$ . Finally, note that **P1-2** is a GP problem with 6K + 1 variables and 10K inequality constraints.

#### IV. OPTIMIZATION ALGORITHM

Although GP problems are not in general convex, they can be transformed straightforwardly to convex problems [49]. For P1-1 and P1-2, it is just a matter of changing each variable "var" by ln("var") and taking the logarithm of the constraint functions (i.e., the posynomials are transformed into log-sum-exp functions, which are convex [49]). After such transformation, each subproblem can be solved by any convex optimization algorithm, by taking advantage of the KKT conditions.

Algorithm 1 details the steps for solving the general optimization problem P1 through the two proposed subproblems. Specifically, lines 1–3 deal with initialization, while lines 4–9 deal with the process of solving **P1-1** and **P1-2** consecutively until the objective function at each iteration decreases at most by  $\xi$ , which is a convergence parameter given as input to the algorithm. Note that after solving P1-1,

which uses (18) as an approximation for (10), we project  $\theta_R$ back to the edge of the original constraint as captured in line 7. After the approximate GP subproblems have been solved, we can still refine (at least some) of the optimization variables to reduce further the objective function. Note that this may be possible since the optimization problem determined by the two GP subproblems operates over a feasible set that is a subset of the original feasible set given by (9b) and (9c). We perform this in line 10 of the optimization algorithm by solving P1-3, which is nothing but **P1** with a fixed input  $\{h, \theta_B, x_{uav}, y_{uav}\}$ and optimization variable set  $\{p_k\}$  as given next

**P1-3:** minimize 
$$t$$
 (27a)

subject to 
$$c_i p_i \le t$$
 (27b)

$$p_i \le p_{\text{max}} \tag{27c}$$

$$-p_i \le p_{\min} \tag{27d}$$

$$\gamma_0 \sum_{j \in \mathcal{K} \setminus i} c_j G_j \bar{g}_j p_j - G_i \bar{g}_i p_i + \sigma^2 \le 0 \quad (27e)$$

where the constraints are  $\forall i \in \mathcal{K}$  and (27e) comes form writing (9b) as a linear equation in  $p_i$  and  $p_i$ . Note that **P1-3** is written in linear programming (LP) form, and since no approximation was used, we can claim global optimality for this particular convex subproblem.

In practice, Algorithm 1 iteratively optimizes the UAV's (position and antenna beamwidth), and the IoT devices' (transmit power) parameters to minimize the energy consumption of the latter. The optimization relies on the known average channel statistics, which depend on the average propagation parameters for different environments, and also on the estimated activation probabilities and relative positions of the IoT devices. Therefore, it can be carried out at the UAV, which periodically reports the updated decisions to the IoT devices. Hence, our proposed solution can be implemented in real-time as long as the system dynamics hold within the convergence time of Algorithm 1. Finally, note that all simplifications that lead to the Algorithm 1 aim at finding a good local optimum of **P1**, for which a convex equivalent form does not exist to the best of our knowledge and, hence, its global optimum cannot be guaranteed by any solver.

Using barrier-based IPM, each GP subproblem in Algorithm 1 can be efficiently solved with accuracy error  $\epsilon$  in a worst case polynomial-time complexity [50]. The number of per-GP required iterations is in the order of

$$C_1 = \mathcal{O}\left(\sqrt{n+m}\ln\frac{(n+m)\Delta}{\epsilon}\right) \tag{28}$$

while each iteration demands

$$C_2 = \mathcal{O}\left((m+s)(s+n)\sqrt{m+n}\ln\frac{(m+n)\Delta}{\epsilon}\right)$$
 (29)

arithmetical operations, where m, n, and s denote the number of constraints, monomial terms, and variables, respectively [51]. The term  $\Delta$  is related to a perturbation in the feasible set when solving the problem. Without loss of generality, we may assume  $\Delta = 1$  since the complexity scales linearly with  $\ln(\Delta)$ . Finally, **P1-3** requires  $\mathcal{O}(\sqrt{m}\log(1/\epsilon))$  iterations each computed in  $\mathcal{O}(s^2m)$  arithmetic operations when using

TABLE II VALUES OF m, n, and s for Each Optimization Subproblem

Parameter	P1-1	P1-2	P1-3
m	5K + 3	10K	4K
n	$K^2 + 4K + 3$	$K^2 + 9K$	_
s	2K+3	6K + 1	K+1

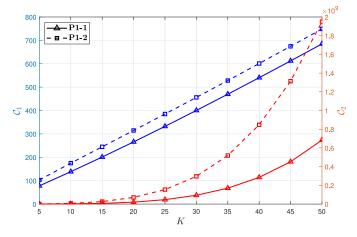


Fig. 2. Complexity analysis versus K of both **P1-1** and **P1-2**, for  $\epsilon = 10^{-2}$ .

the same IPM [52]. Table II presents the values of m, n, and s for each individual GP problem. Fig. 2 shows that as K increases, the number of iterations for solving the GPs grows linearly, whereas the number of operations grows exponentially. The readers can observe that **P1-2** is the most costly subproblem in Algorithm 1, so it roughly determines the final computational time.

## V. NUMERICAL ANALYSIS

In this section, we elucidate numerical results regarding the solution of **P1** throughout three methods. The main approach applies Algorithm 1, where each optimization subproblem is solved with the help of the specialized MOSEK solver [53]. As a benchmark, we solve directly **P1** using IPMs based on logarithmic barrier function and GA. The former solves a sequence of approximate minimization problems by either solving the KKT system or using the conjugate gradient method [54], whereas the latter relies on stochastic derivative-free techniques that mimic the evolution of living species[55].

As Fig. 3 depicts, we consider both: 1) a cellular-like deterministic deployment and 2) random deployments. In both cases the devices are uniformly distributed in the area, although deterministically/randomly in case of deployment 1)/2). Additionally, we assume that the activation probabilities of devices are uniformly distributed such that  $c_k \sim \mathcal{U}(0, 0.5) \ \forall k \in \mathcal{K}$ , regardless of the deployment strategy. In order to study our problem for different propagation conditions, we consider different urban scenarios, namely, suburban, urban, dense urban, and high-rise urban, for which the set of parameters  $\{\psi, \beta, \eta_1, \eta_2\}$  is given in Table III. Finally, unless we state the contrary, the numerical simulations are based on the parameters listed in Table IV. Note that the maximum transmit power meets the typical values of low-power transceivers, e.g., [56], whereas the value of  $\gamma_0$  corresponds

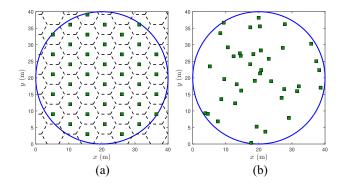


Fig. 3. Analyzed scenarios: (a) deterministic and (b) a random deployment. The green squares represent the IoT devices, whereas the coverage area is delimited with the blue circumference. We set R = 20 m and K = 40.

TABLE III
PROPAGATION PARAMETERS FOR DIFFERENT ENVIRONMENTS [45]

Environment	Parameters $\{\psi, \beta, \eta_1[dB], \eta_2[dB]\}$
Suburban	$\{4.88, 0.43, 0.1, 21\}$
Urban	$\{9.61, 0.16, 1, 20\}$
Dense Urban	{12.08, 0.11, 1.6, 23}
Highrise	{27.23, 0.08, 2.3, 34}

TABLE IV
DEFAULT SIMULATION PARAMETERS

Parameter	Value	Parameter	Value
$\overline{R}$	20 m	$p_{\min}$	1 mW
$\gamma_0$	-16 dB	$p_{ m max}$	500  mW
$ heta_{ m O}$	$\frac{\pi}{18}$	$h_{\min}$	40 m
$f_c$	$2.5{}^{1}\!\!\mathrm{GHz}$	$h_{\max}$	1000 m

to low-rate transmissions as typical in IoT devices. Besides, the carrier frequency matches with the numerical setup in [45] and the hovering height limits lie within the allowed range for a LAP [57].

## A. On the Impact of the Target $\overline{SINR}$

Fig. 4(a) illustrates the objective function as a function of the target  $\overline{SINR}$   $\gamma_0$  for a scenario, where 25 and 50 IoT devices are deployed. Note that the maximum  $\gamma_0$  is subject to Proposition 1's result (12), and it strongly depends on the number of IoT devices. Although the problem was solved using the three previously mentioned optimization tools; Fig. 4(a) shows only the solution provided by Algorithm 1 since IPMs and GA did not often return feasible solutions when varying  $\gamma_0$ . We can note the benefits of a deterministic homogeneous deployment enabled via proper network planning in terms of reducing the average energy consumption at the most energydemanding IoT device. However, the gain with respect to random deployments becomes small as the number of IoT devices decreases; thus, it is not worth restricting their positions; instead, random deployments can provide a quite similar performance with a slight increase of the worst case average energy consumption. Meanwhile, Fig. 4(b) shows the optimal hovering heights of the UAV when 25 devices are randomly and deterministically deployed. As we can notice, lower altitudes are available for the UAV in deterministic deployments when using Algorithm 1 and IPMs. Height results under IPMs exhibit a decreasing behavior with  $\gamma_0$ , while GA does not show a monotonic behavior with the target SINR and Algorithm 1

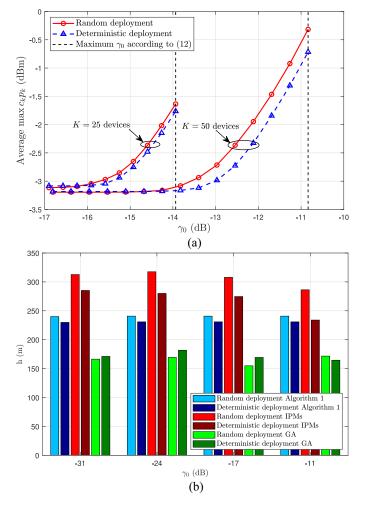


Fig. 4. (a) Average  $\max_k c_k p_k$  for  $K \in \{25, 50\}$  and (b) UAV hovering height for K = 25, versus  $\gamma_0$ , in a dense urban environment using Algorithm 1.

tends to fix the height. The latter may benefit the UAV by saving valuable energy from propelling power.

#### B. On the Impact of the Number of IoT Devices

The impact of the number of devices on the worst case average energy consumption is depicted in Fig. 5. Clearly, Algorithm 1 solution outperforms the other methods in terms of the objective function value, whereas IPMs exhibit some irregularities mainly caused by the nonderivable objective and the highly nonlinear constraint (9b). The worst case average energy consumption has an increasing behavior as a function of the number of IoT devices K. This is because more energy is needed to overcome the interference, which results from the higher number of devices when the network becomes more dense. Herein, the difference of the solution for the deterministic and the random deployments is negligible, showing that there is no additional gain in carefully planning the network. From the results in Fig. 5, we can estimate the battery life of the worst IoT device for each algorithm, which is a common metric when evaluating the network performance. Assuming a low-dense scenario with K = 10 IoT devices equipped with a fully charged battery (for instance, CR MULTICOMP 2032, 3 V, 210 mAh), we have that on average the battery life of the fist device running out of energy is 8 h for IPMs, one day

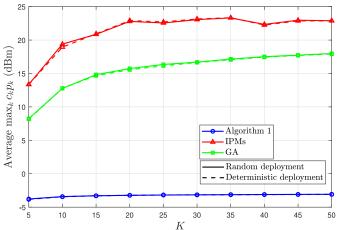


Fig. 5. Average  $\max_k c_k p_k$  for both deterministic and random deployment as a function of the number of IoT nodes K in a dense urban environment.

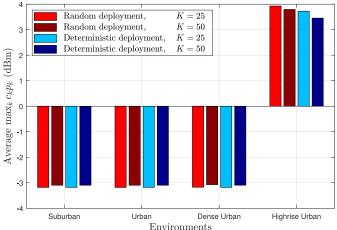


Fig. 6. Average  $\max_k c_k p_k$  evaluation through Algorithm 1 for different urban environments and  $K \in \{25, 50\}$ . The optimum height is approximately 236 m for all scenarios.

and 11 h for GA and 65 days and 22 h for Algorithm 1. The reader can notice that our strategy extends the devices' battery lifetime by minimizing the peaks in their energy consumption profile.

#### C. On the Impact of the Environment and Coverage Area

In Fig. 6, we show the devices' worst case average energy consumption for different urban environments. In the first three environments, the gap among the plots is just appreciated for different number of devices, and not because of the propagation conditions. However, the high-rise environment pushes the energy consumption to larger values to overcome the blockage caused by surrounding objects. In all the cases, the UAV's altitude, which is not shown in the figure, remains nearly constant, and the main distinction is on the transmit power allocation.

On the other hand, the impact of the coverage region dimensions is presented in Fig. 7(a) assuming a dense urban environment. Note that for small coverage areas, the performance remains steady but as the radius increases from certain point (R > 20 m) and R > 25 m for K = 50 and K = 25, respectively), all curves start to steadily increase. This behavior is supported by the fact that the UAV goes higher in order to

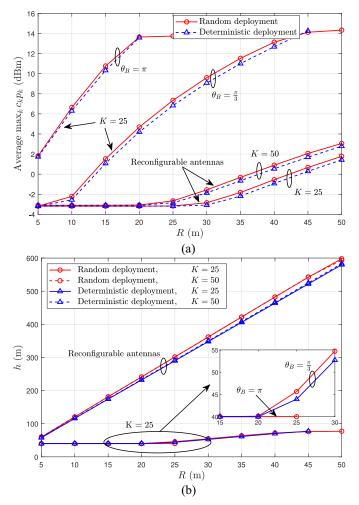


Fig. 7. (a) Average  $\max_k c_k p_k$  and (b) optimal UAV hovering height versus R. We consider a dense urban environment and  $K \in \{25, 50\}$ . The attained optimum beamwidth remains constant over R:  $\theta_R^* = \theta_0 = (\pi/18)$ .

serve larger areas, as shown in Fig. 7(b), which demands greater energy resources from the IoT devices. We have shown the results delivered by Algorithm 1, since IPMs and GA do not converge smoothly to the final solution. Note that for  $R \ge 30$  m, the UAV requires to fly 15-m higher on average, to serve the IoT devices in the random deployment over the case when the network has been planned, which costs a slight increase in the UAV's propelling power. As a benchmark, we also present the results of using antennas with fixed beamwidth at the UAV. The reader can notice the performance degradation in terms of worst case average energy consumption as  $\theta_B$ increases [Fig. 7(a)], despite the fact that the hovering height has decreased [Fig. 7(b)] with respect to the scenario with reconfigurable antennas. This is because the antenna gain has significantly deteriorated and the IoT devices must transmit with higher power. In particular, the optimization problem is no longer feasible for R > 25 m, when  $\theta_B = \pi$ . We can also notice that the optimal UAV hovering heights in both Figs. 4(b) and 7(b) fall within the allowed range for the operation of LAPs [58].

Fig. 8 shows the average computation time (using parallel quad-core processing) for all methods as a function of the number of devices. As observed, the average computation

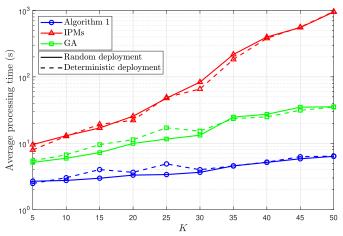


Fig. 8. Average computation time versus K in a dense urban environment.

time grows exponentially as a function of the number of devices K when using IPMs and GA methods, while the increase is closely linear when using Algorithm 1, which is expected according to our discussions around Fig. 2. Note that Algorithm 1 is the least time-consuming method for every K—it converges in around three iterations—while it is worth mentioning that IPMs have the highest computation time because of the extreme nonlinearity of (9b), which does not impact significantly the other approaches. Finally, computing the solution under random deployment costs slightly less time when compared to deterministic deployment.

#### VI. CONCLUSION

In this work, we proposed a GP-based algorithm for minimizing the worst case average energy consumption of IoT devices served by a hovering UAV. The proposed algorithm not only controls the uplink transmit power of the devices but also the UAV's position and antenna configuration. We quantified the maximum attainable QoS in terms of perlink SINR as a function of the number of IoT devices and their activation probabilities, and showed how it diminishes in crowded deployments due to excessive interference. Our proposed optimization algorithm stands out when compared to two different benchmark schemes relying on IPMs and GA solvers, with clear reduction in the computer processing time. Additionally, we showed the marginal benefits of planning the network compared to the random deployment in terms of reducing the worst case average energy consumption, especially in dense deployments. Among the analyzed urban environments, the high-rise environment was shown to be the most demanding in terms of energy consumption. Finally, the worst case average energy requirements were shown to increase as the UAV flies higher to serve wider areas.

As an interesting future work, we could also consider energy consumption and maximum flying speed at the UAV, which are limiting factors of UAV-assisted networks [59]. Besides, we can include specific regulations on the maximum hovering altitude in the worst case average energy minimization problem. Finally, the trajectory of a UAVs' swarm can be optimized for serving the IoT network as an extension for this work instead of utilizing a single UAV.

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Osmel Martínez Rosabal was born in Camagüey, Cuba, in 1989. He received the B.Sc. degree (with distinction) in telecommunications and electronic engineering from the Central University of Las Villas, Santa Clara, Cuba, in 2013, and the M.Sc. degree (with distinction) in wireless communications engineering from the University of Oulu, Oulu, Finland, in 2021, where he is currently pursuing the Doctoral degree with the Centre for Wireless Communications.

From 2013 to 2018, he served as a Specialist in Telematics with the Cuban Telecommunication Company, Sancti Spiritus, Cuba. His research interests include energy harvesting, sustainable wireless energy transfer, machine-type communications, and green IoT.



Onel Luis Alcaraz López (Member, IEEE) was born in Sancti Spiritus, Cuba, in 1989. He received the B.Sc. degree (1st class Hons.) in electrical engineering from the Central University of Las Villas, Santa Clara, Cuba, in 2013, the M.Sc. degree in electrical engineering from the Federal University of Paraná, Curitiba, Brazil, in 2017, and the D.Sc. degree (with distinction) in electrical engineering from the University of Oulu, Oulu, Finland, in 2020.

From 2013 to 2015, he served as a Specialist in Telematics with the Cuban Telecommunications

Company, Sancti Spiritus, Cuba. He has coauthored the book entitled *Wireless RF Energy Transfer in the Massive IoT Era: Towards Sustainable Zero-Energy Networks*, (Wiley, Dec 2021). He currently holds an Assistant Professorship (Tenure Track) in Sustainable Wireless Communications Engineering with the Centre for Wireless Communications, University of Oulu. His research interests include sustainable IoT, energy harvesting, wireless RF energy transfer, wireless connectivity, machine-type communications, and cellular-enabled positioning systems.

Dr. Luis Alcaraz López is a Collaborator to the 2016 Research Award given by the Cuban Academy of Sciences, a co-recipient of the 2019 IEEE European Conference on Networks and Communications Best Student Paper Award, and the recipient of the 2020 Best Doctoral Thesis Award granted by Academic Engineers and Architects in Finland TEK and Tekniska Föreningen i Finland TFiF in 2021.



Dian Echevarría Pérez (Graduate Student Member, IEEE) was born in Cienfuegos, Cuba, in 1989. He received the B.Sc. degree in telecommunications and electronic engineering from the Central University of Las Villas, Santa Clara, Cuba, in 2013, and the M.Sc. degree (with distinction) in wireless communications engineering from the University of Oulu, Finland, in 2021, where he is currently pursuing the Doctoral degree with the Centre for Wireless Communications.

From 2013 to 2018, he worked as a Specialist in

Telematics with the Cuban Telecommunications Company, Sancti Spiritus, Cuba. His research interests are full-duplex communications, wireless RF energy transfer, ultrareliable low-latency communications, and resource allocation in schemes with partial/limited channel state information.



Mohammad Shehab received the B.Sc. degree from Alexandria University, Alexandria, Egypt, in 2011, the M.Sc. degree from the Arab Academy, Alexandria, in 2014, the M.Sc. degree from the University of Oulu, Oulu, Finland, in 2017, and the Doctoral degree from the University of Oulu in 2022, with focus on energy-efficient QoS provisioning in IoT

He worked as a TA with Alexandria University and the Arab Academy from 2012 to 2015. His work so far resulted in ten conference papers, six journal

papers, and one patent.

Dr. Shehab won the Best Student Paper Award in ISWCS 2017, and the Nokia Foundation Award consecutively for 2018 and 2019.



Henrique Hilleshein was born in Santo Amaro da Imperatriz, Santa Catarina, Brazil, in October 11, 1995. He received the Bachelor of Science degree in telecommunications engineering from the Federal Institute of Santa Catarina, Florianópolis, Brazil, in 2018, and the Master of Science degree in wireless communications engineering from the University of Oulu, Oulu, Finland, in 2021.

In 2019, he joined the Machine-Type Wireless Communications Group, Centre for Wireless Communications, University of Oulu, as a Research

Assistant. Since his master's degree graduation, he has been working outside of academia.



Hirley Alves (Member, IEEE) received the B.Sc. and M.Sc. degrees in electrical engineering from the Federal University of Technology-Paraná (UTFPR), Curitiba, Brazil, in 2010 and 2011, respectively, and the dual D.Sc. degree from the University of Oulu, Oulu, Finland, and UTFPR, in 2015.

In 2017, he was an Adjunct Professor in machinetype wireless communications with the Centre for Wireless Communications (CWC), University of Oulu. In 2019, he joined CWC as an Assistant Professor and is currently the Head of the Machine-

type Wireless Communications Group. He is actively working on massive connectivity and ultra-reliable low latency communications for future wireless networks, 5GB and 6G, full-duplex communications, and physical-layer security. He leads the URLLC activities for the 6G Flagship Program.

Prof. Alves is a co-recipient of the 2017 IEEE International Symposium on Wireless Communications and Systems (ISWCS) Best Student Paper Award and the 2019 IEEE European Conference on Networks and Communications (EuCNC) Best Student Paper Award co-recipient of the 2016 Research Award from the Cuban Academy of Sciences. He has been the organizer, chair, TPC, and tutorial lecturer for several renowned international conferences. He is the General Chair of the ISWCS'2019 and the General Co-Chair of the 1st 6G Summit, Levi 2019, and ISWCS 2020.